

**ROYAL MILITARY COLLEGE OF CANADA**  
**FALL TERM EXAMINATIONS 2018 – 2019**  
**PG FINAL EXAMINATION**

**EE501: INTRODUCTION TO THE THEORY OF STATISTICAL COMMUNICATION**

**Monday, 17 December 2018, 1300 - 1600 h.**

**EXAMINER:** G Drolet, Associate Professor

**TIME:** 3 hours

- NOTES:**
1. Hand held calculator is allowed.
  2. Only the textbook (Wozencraft & Jacobs) and course notes are allowed; problem solutions are not allowed.
  3. Answer only two of questions 1, 2, 3. Answer all of question 4.
  4. The questions have the following value:
    - Question #1, 5 points
    - Question #2, 5 points
    - Question #3: 5 points
    - Question #4: 15 points
  5. Justify all your answers.

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Question 1

Answer only two of questions 1, 2, 3.

- <sup>1</sup> A communication system is used to transmit one of two equally likely messages,  $m_0$  and  $m_1$ . The channel output is a continuous random variable  $r$ , the conditional density functions of which are shown in Fig. 1. Determine the optimum receiver decision rule and compute the resulting probability of error.

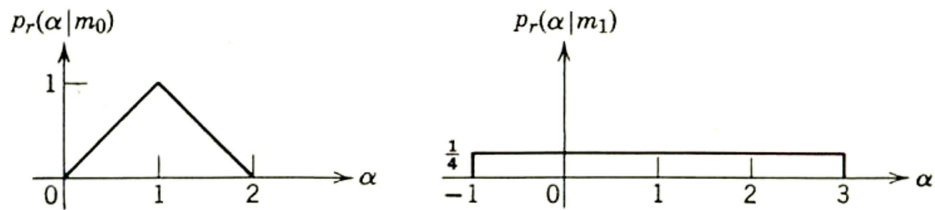


Figure 1:

- <sup>2</sup> It is known that  $P[\mathcal{E}]_{\min} = q$  when the two signal vectors  $\mathbf{s}_0$  and  $\mathbf{s}_1$  shown in fig 2a are transmitted with equal probability over a channel disturbed by additive white Gaussian noise. Compute  $P[\mathcal{E}]_{\min}$  in terms of  $q$ ,  $\theta$ , and  $l$  when the nine vectors indicated by  $\times$ 's in fig 2b are used as signals with equal probability over the same channel.

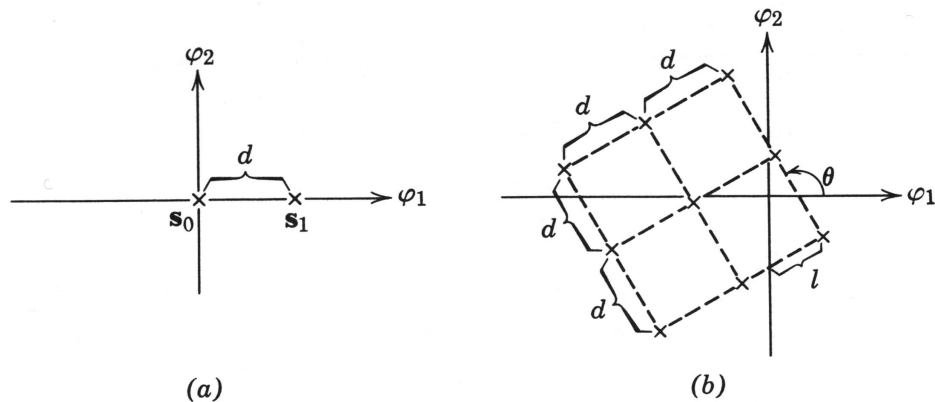


Figure 2:

<sup>1</sup>Problem 2.23 in the textbook by Wozencraft & Jacobs.

<sup>2</sup>Problem 4.3, pages 274 in W & J.

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Question 3

3. In the diagram of figure 3,  $n_w(t)$  denotes a white Gaussian noise with power spectral density  $S_w(f) = \frac{\mathcal{N}_0}{2}$  and  $r = y(0)$ . Express  $\bar{r}$ ,  $Var(r)$  as a function of  $\mathcal{N}_0/2$ .

**Hint:** You may interchange  $E[\ ]$  and  $\int$  and we recall that:

$$\begin{aligned} f(t) * g(t) &= g(t) * f(t) \\ &= \int_{-\infty}^{\infty} f(\alpha)g(t - \alpha)d\alpha \\ &= \int_{-\infty}^{\infty} g(\alpha)f(t - \alpha)d\alpha \end{aligned}$$

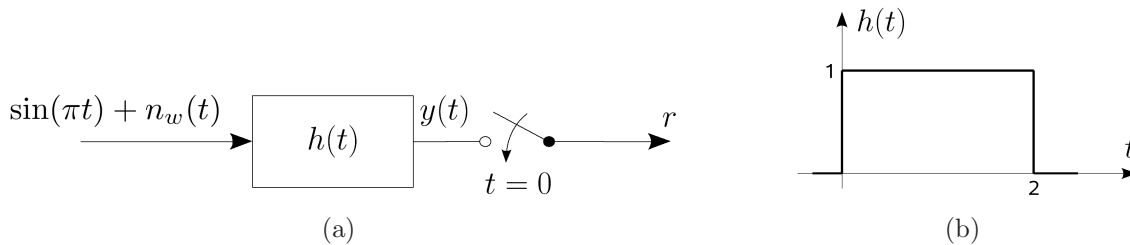


Figure 3:

Answer all of question 4.

4. The three signals  $s_0(t)$ ,  $s_1(t)$ ,  $s_2(t)$  given by<sup>3</sup>:

$$\begin{aligned} s_0(t) &= (u(t) - u(t - 1))/2 \\ s_1(t) &= t(u(t) - u(t - 1)) \\ s_2(t) &= \cos(2\pi t)(u(t) - u(t - 1)) \end{aligned}$$

are used to communicate one of three equally likely messages on a channel with additive white Gaussian noise  $n_w(t)$  statistically independent of the message and with power spectral density  $\mathcal{N}_0/2$ .  $s_0(t)$ ,  $s_1(t)$ ,  $s_2(t)$  are sketched in figure 4 and their energy are easily found (you need not verify this):

$$\begin{aligned} E_{s_0} &= 1/4 \\ E_{s_1} &= 1/3 \\ E_{s_2} &= 1/2 \end{aligned}$$

The following is an orthonormal basis suitable for the representation of signals  $s_0(t)$ ,  $s_1(t)$  (you need not verify this):

$$\begin{aligned} \phi_1(t) &= \sqrt{3}(1 - t)(u(t) - u(t - 1)) \\ \phi_2(t) &= (3t - 1)(u(t) - u(t - 1)) \end{aligned}$$

$\phi_1(t)$ ,  $\phi_2(t)$  are sketched in figure 5.

<sup>3</sup>  $u(t)$  denotes as usual the unit step function.

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Question 4

- (a) Calculate a basis function  $\phi_3(t)$ , if required, which, together with  $\phi_1(t)$ ,  $\phi_2(t)$ , forms an orthonormal basis suitable for the representation of all three signals  $s_0(t)$ ,  $s_1(t)$  and  $s_2(t)$ .

**Note:** If you are unable to calculate  $\phi_3(t)$ , you may use the basis function  $\phi'_3(t)$  given by:

$$\phi'_3(t) = \sqrt{\frac{5\pi^2}{180 + 91\pi^2}} (6t^2 + 6 \cos(2\pi t) - 6t + 1)$$

in the subquestions that follow.  $\{\phi_1(t), \phi_2(t), \phi'_3(t)\}$  forms an orthonormal basis that spans  $s_0(t)$ ,  $s_1(t)$  but not  $s_2(t)$ ; it may nonetheless be used in the remainder of this problem.

- (b) Find vector representation of signals  $s_0(t)$ ,  $s_1(t)$ ,  $s_2(t)$  in the basis  $\phi_1(t)$ ,  $\phi_2(t)$ ,  $\phi_3(t)$  (i.e. as triplets).

**Hint:** Some reflection may save you a lot of time; think before you start some long and tedious calculations.

**Note:** If you are using  $\phi'_3(t)$ , then you would find:

$$\langle s_2(t), \phi'_3(t) \rangle = \frac{3(1 + \pi^2)}{\pi} \sqrt{\frac{5}{180 + 91\pi^2}} \approx 0.7068605607$$

Do not calculate this value.

- (c) If the transmitted waveform and the noise are such that the received waveform is

$$r(t) = s(t) + n_w(t) = \frac{4}{5} (10t^2 - 10t + 1)$$

which of the three signals  $s_0(t)$ ,  $s_1(t)$ ,  $s_2(t)$  was most likely transmitted.

**Hint:** Again some reflection may save you a lot of time; choose carefully which receiver implementation to use.

- (d) What is the average energy per message and what value of  $\mathcal{N}_0$  corresponds to a signal-to-noise ratio of 15 dB.

**Note:** If you can't calculate the value of  $\mathcal{N}_0$  you may use  $\mathcal{N}_0 = 0.01$  in the next subquestion.

- (e) Use the union bound to obtain an upper bound on the probability of error of the communication system at a signal to noise ratio of 15 dB.

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Question 4

The following may be useful:

$$\sqrt{\langle \mathbf{u} - \mathbf{v}, \mathbf{u} - \mathbf{v} \rangle} \equiv \text{distance between } \mathbf{u}, \mathbf{v}$$

$$\int t^n dt = \frac{t^{n+1}}{n+1}$$

$$\int \frac{1}{t} dt = \ln(t)$$

$$\int \cos(2\pi t) dt = \frac{\sin(2\pi t)}{2\pi}$$

$$\int \cos^2(2\pi t) dt = \frac{\cos(2\pi t) \sin(2\pi t) + 2\pi t}{4\pi}$$

$$\int t \cos(2\pi t) dt = \frac{\cos(2\pi t) + 2t \sin(2\pi t) \pi}{4\pi^2}$$

$$\int t^2 \cos(2\pi t) dt = \frac{2\pi^2 t^2 \sin(2\pi t) - \sin(2\pi t) + 2\pi t \cos(2\pi t)}{4\pi^3}$$

$$\int t^3 \cos(2\pi t) dt =$$

$$\frac{4\pi^3 t^3 \sin(2\pi t) + 6\pi^2 t^2 \cos(2\pi t) - 3 \cos(2\pi t) - 6t \sin(2\pi t) \pi}{8\pi^4}$$

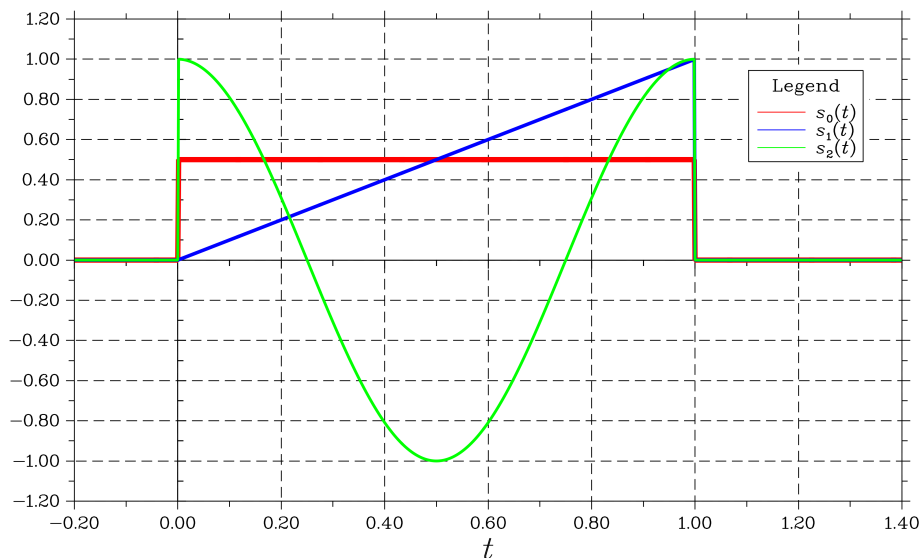


Figure 4:

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Question 4

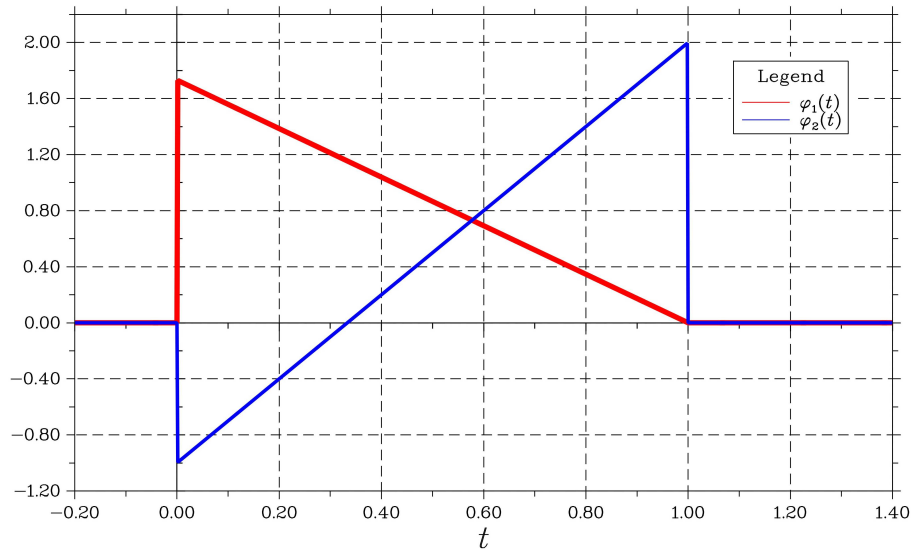


Figure 5:

END

**Table of the  $Q(\ )$  and  $\text{erf}(\ )$  functions**

The approximation  $Q(x) \approx \frac{1}{x\sqrt{2\pi}}(1 - \frac{0.7}{x^2})e^{-x^2/2}$  may be used when  $x > 2$ .

$x$	$\text{erf}(x)$	$Q(x)$	$x$	$\text{erf}(x)$	$Q(x)$	$x$	$\text{erf}(x)$	$Q(x)$
0.00	0	0.5	1.70	0.9838	0.04457	3.40	1	0.0003369
0.10	0.1125	0.4602	1.80	0.9891	0.03593	3.50	1	0.0002326
0.20	0.2227	0.4207	1.90	0.9928	0.02872	3.60	1	0.0001591
0.30	0.3286	0.3821	2.00	0.9953	0.02275	3.70	1	0.0001078
0.40	0.4284	0.3446	2.10	0.997	0.01786	3.80	1	$7.235 \times 10^{-5}$
0.50	0.5205	0.3085	2.20	0.9981	0.0139	3.90	1	$4.810 \times 10^{-5}$
0.60	0.6039	0.2743	2.30	0.9989	0.01072	4.00	1	$3.167 \times 10^{-5}$
0.70	0.6778	0.242	2.40	0.9993	0.008198	4.10	1	$2.066 \times 10^{-5}$
0.80	0.7421	0.2119	2.50	0.9996	0.00621	4.20	1	$1.335 \times 10^{-5}$
0.90	0.7969	0.1841	2.60	0.9998	0.004661	4.30	1	$8.540 \times 10^{-6}$
1.00	0.8427	0.1587	2.70	0.9999	0.003467	4.40	1	$5.413 \times 10^{-6}$
1.10	0.8802	0.1357	2.80	0.9999	0.002555	4.50	1	$3.398 \times 10^{-6}$
1.20	0.9103	0.1151	2.90	1	0.001866	4.60	1	$2.112 \times 10^{-6}$
1.30	0.934	0.0968	3.00	1	0.00135	4.70	1	$1.301 \times 10^{-6}$
1.40	0.9523	0.08076	3.10	1	0.0009676	4.80	1	$7.933 \times 10^{-7}$
1.50	0.9661	0.06681	3.20	1	0.0006871	4.90	1	$4.792 \times 10^{-7}$
1.60	0.9763	0.0548	3.30	1	0.0004834	5.00	1	$2.867 \times 10^{-7}$